

The Johns Hopkins University
Department of Electrical and Computer Engineering

505.460 — Introduction to Linear Systems — Fall 2002

Final exam

Name: _____

You are allowed to use:

1. Table 3.1 (page 206) & Table 3.2 (page 221)
2. Table 4.1 (page 328) & Table 4.2 (page 329)
3. Table 5.1 (page 391) & Table 5.2 (page 392)
4. One standard size, double sided formula sheet.
5. Answer the questions in the sheet below.

1. a b c d e
2. a b c d e
3. a b c d e
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18. a b c d e
19. a b c d e
20. a b c d e

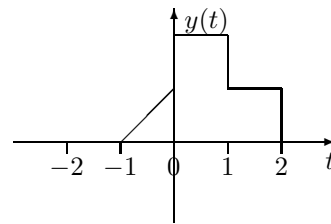
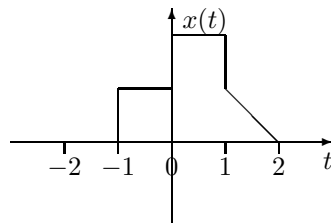
1. Which of the following signals is equivalent to $x[n] = nu[n] - u[n - 2]$?

- (a) $x[n] = \delta[n] + \delta[n - 1]$
- (b) $x[n] = \delta[n] - \delta[n - 2]$
- (c) $x[n] = \delta[n]$
- (d) $x[n] = \delta[n - 1] + (n - 2)u[n - 2]$
- (e) None of the above.

2. Which of the following statements is not true?

- (a) $x(t) = e^{j2t/3}$ is continuous-time periodic.
- (b) $x[n] = e^{2t}u(-t)$ is bounded.
- (c) $x[n] = e^{j2n/3}$ is discrete-time periodic.
- (d) $x[n] = e^{jt} + e^{-jt}$ is periodic.
- (e) None of the above.

3. The signal $x(t)$ is shown below. Also shown is $y(t)$.



The relationship between $y(t)$ and $x(t)$ is:

- (a) $y(t) = x(t - 1)$
- (b) $y(t) = x(2 - t)$
- (c) $y(t) = x(1 - t)$
- (d) $y(t) = x(t - 2)$
- (e) None of the above.

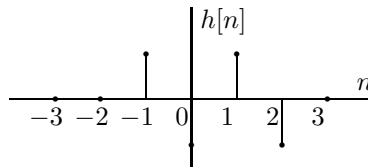
4. The system described by $y(t) = (x(t)x(t - 2))u(t - 1)$ is...

- (a) memoryless, time-invariant and linear.
- (b) time-invariant, linear and causal.
- (c) linear, causal and stable.
- (d) causal, stable and memoryless.
- (e) None of the above.

5. An LTI, discrete-time system has impulse response $h[n] = (5)^n (u[n - 1] + u[1 - n])$; is the system

- (a) stable, but not causal?
- (b) causal, but not stable?
- (c) stable and causal?
- (d) not stable and not causal?
- (e) None of the above.

6. If $h[n] = \delta[n] + \delta[n - 1]$ and $x[n] = nu[n] + \delta[n - 2]$, then
- $y[n] = nu[n] + nu[n - 1] + \delta[n - 1] + \delta[n - 2]$
 - $y[n] = nu[n] + (n - 1)u[n - 2] + \delta[n - 1] + \delta[n - 2]$
 - $y[n] = nu[n] + (n - 1)u[n - 1] + \delta[n - 2] + \delta[n - 3]$
 - $y[n] = nu[n] + nu[n - 2] + \delta[n - 2] + \delta[n - 3]$
 - None of the above.
7. For an LTI system with impulse response $h(t) = -\delta(t) + e^{-t}u(t)$ and input signal $x(t) = u(t)$, the value of the output signal at $t = 1$ is
- $y(1) = 0$.
 - $y(1) = -\infty$.
 - $y(1) = 2 + 1/e$.
 - $y(1) = -1/e$.
 - None of the above.
8. For a discrete-time LTI system with impulse response $h[n] = nu[n]$ and input sketched below,



the output signal $y[n]$ is zero on the range

- $n \leq 0, n \geq 5$.
 - $n \leq -1$.
 - $n \geq 5$.
 - $n \leq -1, n \geq 5$.
 - None of the above.
9. The Fourier series representation of a function has non-zero coefficients $a_0 = 1$, $a_1 = a_{-1}^* = 2j$ and $a_3 = a_{-3} = 1$ and period $T_0 = 2$. The function $x(t)$ is
- $2 - 2 \cos(2\pi t) + 2 \sin(6\pi t)$
 - $1 - 4 \sin(2\pi t) + \sin(6\pi t)$
 - $2 + 2 \cos(\pi t) + \cos(3\pi t)$
 - $1 - 4 \sin(\pi t) + 2 \cos(3\pi t)$
 - None of the above.
10. If $x(t) = 2 \sin\left(\frac{2\pi}{3}t\right) + 4 \cos\left(\frac{5\pi}{3}t\right)$, the fundamental period and Fourier series coefficients are
- $T_0 = 6, a_2 = a_{-2}^* = -j, a_5 = a_{-5} = 2$
 - $T_0 = 3, a_1 = a_{-1}^* = -j, a_5 = a_{-5} = 4$
 - $T_0 = 3, a_2 = a_{-2}^* = -j, a_5 = a_{-5} = 2$
 - $T_0 = 6, a_1 = a_{-1}^* = -j, a_5 = a_{-5} = 2$
 - None of the above.

11. If the continuous-time periodic signal $x(t)$ has a Fourier series representation $x(t) = \sum_{k=-\infty}^{\infty} a_k e^{j\omega_0 kt}$ then the signal

$$y(t) = \frac{1}{3}x(t-1) + \frac{1}{3}x(t) + \frac{1}{3}x(t+1)$$

has Fourier series representation $y(t) = \sum_{k=-\infty}^{\infty} b_k e^{j\omega_0 kt}$ where

- (a) $b_k = \frac{1}{3}a_{k-1} + \frac{1}{3}a_k + \frac{1}{3}a_{k+1}$
 - (b) $b_k = \frac{1}{3}(2 \cos \omega_0 k + 1)a_k$
 - (c) $b_k = \frac{1}{3}a_{k-1}e^{-j\omega_0 t} + \frac{1}{3}a_k + \frac{1}{3}a_{k+1}e^{j\omega_0 t}$
 - (d) $b_k = a_{k-1}e^{-j\omega_0 t} + a_k + a_{k+1}e^{j\omega_0 t}$
 - (e) None of the above.
12. If $X(\omega)$ is the Fourier Transform of $x(t)$, and $z(t) = -3x(10(t-1))$.
- (a) $Z(j\omega) = \frac{3}{10}X(j\frac{\omega}{10})e^{j\omega}$
 - (b) $Z(j\omega) = -\frac{3}{10}X(j\frac{\omega}{10})e^{j\omega}$
 - (c) $Z(j\omega) = \frac{3}{20}X(j\frac{\omega}{5})e^{-j\omega}$
 - (d) $Z(j\omega) = -\frac{3}{20}X(j\frac{\omega}{5})e^{-j\omega}$
 - (e) None of the above.

13. If $x(t) = 4 \sin t \cos t$

- (a) $X(\omega)$ does not exist.
- (b) $X(\omega) = 4\pi\delta(\omega) + 2\pi\delta(\omega-2) + 2\pi\delta(\omega+2)$.
- (c) $X(\omega) = 4\pi\delta(\omega) + \frac{1}{2+j\omega} - \frac{1}{2-j\omega}$.
- (d) $X(\omega) = 4\pi\delta(\omega-2) + 4\pi\delta(\omega+2)$.
- (e) None of the above.

14. For the discrete-time function $x[n] = (\frac{1}{2})^{|n|}$, the discrete-time Fourier transform satisfies

- (a) $\text{Im } X(e^{j\omega}) = 0$.
- (b) $\text{Re } X(e^{j\omega}) = 0$.
- (c) $\int_{-\pi}^{\pi} X(e^{j\omega}) d\omega = 0$
- (d) $X(e^{j0}) = 0$
- (e) None of the above.

15. Ideal high pass filters can not be used in used in real-time applications because:

- (a) They are too expensive.
- (b) They require too much bandwidth.
- (c) They can be made directly from low pass filters.
- (d) They are not causal.
- (e) None of the above.

16. Suppose that $h_{W_1}(t)$ represents the impulse response of an ideal low pass filter with cutoff W_1 Hz. Which of these represents the impulse response of an ideal band-pass filter that passes signals with frequencies between 100 and 200 Hz?
- (a) $(\delta(t) - h_{100}(t)) * h_{200}(t)$
 - (b) $(\delta(t) - h_{200}(t)) * h_{100}(t)$
 - (c) $(1 - h_{200}(t)) * h_{100}(t)$
 - (d) $(1 - h_{100}(t)) * h_{200}(t)$
 - (e) None of the above.
17. For the signal
- $$x(t) = \sin(10\pi t) + 2 \cos(20\pi t)$$
- the minimum sampling frequency that can be used to obtain samples of $x(t)$ without loss of information is
- (a) 10 Hz.
 - (b) 10π Hz.
 - (c) 20 Hz.
 - (d) 40 Hz.
 - (e) None of the above.
18. If the Nyquist frequency for $x_1(t)$ is ω_1 and the Nyquist frequency for $x_2(t)$ is ω_2 , then the Nyquist frequency for the convolution of x_1 and x_2 is
- (a) $\max(\omega_1, \omega_2)$
 - (b) $\min(\omega_1, \omega_2)$
 - (c) $\omega_1 \times \omega_2$
 - (d) $\omega_1 + \omega_2$
 - (e) None of the above.
19. A DSB/SC modulation scheme
- (a) Sends a copy of the carrier.
 - (b) Requires more bandwidth than a SSB scheme.
 - (c) Is ideal for a situation where there is one sender and many receivers, such as commercial radio.
 - (d) All of the above.
 - (e) None of the above.
20. This teaching in this course has been...
- (a) Fascinating.
 - (b) Scintillating.
 - (c) Lucid.
 - (d) The best I have ever seen.
 - (e) All of the above.